

## UNIT 2

### Exercises of UNIT 2- Part II: Spectral Characteristics

2.10

Given that  $X(t) = \sum_{i=1}^N \alpha_i X_i(t)$  where  $\alpha_i$  are real constants, show that

$$S_{XX}(\omega) = \sum_{i=1}^N \alpha_i^2 S_{X_i X_i}(\omega)$$

if

- (a) the processes  $X_i(t)$  are orthogonal
- (b) the processes are independent with zero mean

a)  $S_{XX}(\omega)$ ?

$X_i, X_j$  are orthogonal  $\Rightarrow E[X_i(t_1) X_j(t_2)] = 0$   
 $(i \neq j)$   $R_{X_i X_j}(t_1, t_2) = 0$

$$S_{XX}(\omega) = FT \langle R_{XX}(t) \rangle$$

$$R_{XX}(t_1, t_2) = E[X(t_1) X(t_2)] = E \left[ \sum_{i=1}^N \alpha_i X_i(t_1) \sum_{j=1}^N \alpha_j X_j(t_2) \right]$$

$$X(t) = \sum_{i=1}^N \alpha_i X_i(t)$$

$$= E \left[ \sum_{i=1}^N \sum_{j=1}^N \alpha_i \alpha_j X_i(t_1) X_j(t_2) \right]$$

$$= \sum_{i=1}^N \sum_{j=1}^N \alpha_i \alpha_j E[X_i(t_1) X_j(t_2)]$$

$$= \underbrace{\sum_i \alpha_i^2 E[X_i(t_1) X_i(t_2)]}_{R_{X_i X_i}(t_1, t_2)} + \sum_{\substack{i=1 \\ i \neq j}}^N \sum_{j=1}^N \alpha_i \alpha_j \underbrace{E[X_i(t_1) X_j(t_2)]}_{R_{X_i X_j}(t_1, t_2) = 0}$$

$$= \sum_i \alpha_i^2 R_{X_i X_i}(t_1, t_2) = R_{XX}(t_1, t_2) \quad \underline{X_i, X_j \text{ are orthogonal}}$$

$$= \sum_i \alpha_i^2 R_{X_i X_i}(\tau) \quad \leftarrow \text{LET'S ASSUME THIS}$$

$$\begin{aligned}
 S_{XX}(\omega) &= \text{FT} \langle R_{XX}(\tau) \rangle = \text{FT} \langle \sum_i \alpha_i^2 R_{X_i X_i}(\tau) \rangle \\
 &= \sum_{i=1}^N \alpha_i^2 \underbrace{\text{FT} \langle R_{X_i X_i}(\tau) \rangle}_{S_{X_i X_i}(\omega)} = \sum_{i=1}^N \alpha_i^2 S_{X_i X_i}(\omega)
 \end{aligned}$$

b)  $S_{XX}(\omega)$ ?

$x_i, x_j$  INDEPENDENT  
 $i \neq j$

$\forall i, E[x_i(t)] = 0$

$\text{IND} \rightarrow E[x_i(t_i)x_j(t_i)] = E[x_i(t_i)]E[x_j(t_i)] = 0$

from mixed orthogonal

2.11

If  $X(t)$  is a stationary process, find the power spectrum of

$Y(t) = A_0 + B_0X(t)$  in terms of the power spectrum of  $X(t)$  if  $A_0$  and  $B_0$  are real constants.

2.12 The autocorrelation function of a random process  $X(t)$  is

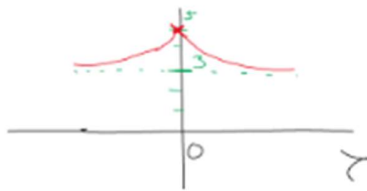
$$R_{XX}(\tau) = 3 + 2e^{-4\tau}$$

Find

- (a) The power spectrum of  $X(t)$
- (b) The average power of  $X(t)$

(c) The fraction of power that lies in the frequency band

$$-\sqrt{\frac{1}{2}} \leq \omega \leq \sqrt{1/2}$$



2       $\alpha/2\pi$

$\alpha\delta(\omega)$

20       $e^{-\tau^2/(2\sigma^2)}$

$\sigma\sqrt{2\pi}e^{-\sigma^2\omega^2/2}$

$$P_{xx}(\tau) = 3 + 2e^{-\tau^2}$$

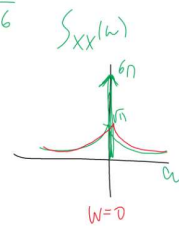
a)  $S_{xx}(\omega)$ ?

$$S_{xx}(\omega) = \text{FT}\langle R_{xx}(\tau) \rangle$$

$$= \text{FT}\langle 3 \rangle + \text{FT}\langle 2e^{-4\tau^2} \rangle$$

$$= 2\pi \cdot 3 \delta(\omega) + \frac{2}{\sqrt{8}} e^{-\frac{\omega^2}{16}}$$

$$= 6\pi \delta(\omega) + \sqrt{\pi} e^{-\frac{\omega^2}{16}}$$



b)  $P_{xx}$ ?

$$P_{xx} = R_{xx}(0) = 3 + 2 = 5 \text{ W}$$

$$P_{xx} = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_{xx}(\omega) d\omega = \frac{1}{2\pi} \left[ \underbrace{\int_{-\infty}^{\infty} 6\pi \delta(\omega) d\omega}_{6\pi} + \int_{-\infty}^{\infty} \sqrt{\pi} e^{-\frac{\omega^2}{16}} d\omega \right]$$



(a)  $S_{XX}(\omega)$

(b)  $P_{XX}$  from  $S_{XX}(\omega)$

(c)  $P_{XX}$  from  $R_{XX}(\tau)$

TIP: The clue for this exercise is to express  $\cos^4$

as a summation of cosine functions:

$$\begin{aligned}\cos^4(\omega\tau) &= \cos^2(\omega\tau) \cdot \cos^2(\omega\tau) \\ &= \frac{1}{2}(1 + \cos(2\omega\tau))(1 + \cos(2\omega\tau)) \\ &= \frac{1}{2}(1 + \cos(2\omega\tau) + \cos(2\omega\tau) + \cos^2(2\omega\tau)) \\ &= \frac{1}{2}(1 + 2\cos(2\omega\tau) + 1 + \cos(4\omega\tau)) \\ &= 1 + \cos(2\omega\tau) + \frac{1}{2}\cos(4\omega\tau)\end{aligned}$$

2.14 Given a random process with autocorrelation

$R_{XX}(\tau) = Ae^{-\alpha|\tau|}\cos(\omega_0\tau)$  where  $A > 0$ ,  $\alpha > 0$ , and  $\omega_0$  are real constants, find the power spectrum.

$$R_{XX}(z) = A e^{-\alpha|z|} \cos(\omega_0 z)$$

$A, \omega_0 \rightarrow \text{constant}$

$$\begin{aligned} S_{XX}(\omega) &= \text{FT} \langle R_{XX}(z) \rangle = \text{FT} \langle A e^{-\alpha|z|} \cos \omega_0 z \rangle \\ &= A \left[ \underbrace{\text{FT} \langle e^{-\alpha|z|} \rangle}_{\frac{2\alpha}{\alpha^2 + \omega^2}} * \underbrace{\text{FT} \langle \cos \omega_0 z \rangle}_{\pi [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]} \right] \\ &= A\pi \cdot \left( \frac{2\alpha}{\alpha^2 + \omega^2} \right) * [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] \\ &= 2\pi\alpha A \left[ \frac{1}{\alpha^2 + (\omega - \omega_0)^2} + \frac{1}{\alpha^2 + (\omega + \omega_0)^2} \right] \end{aligned}$$

2.15 A random process is given by

$W(t) = AX(t) + BY(t)$  where  $A$  and  $B$  are real constants and  $X(t)$  and  $Y(t)$  are jointly wide-sense stationary processes. Find

- The power spectrum of  $W(t)$  as a function of  $S_{XX}(\omega)$ ,  $S_{YY}(\omega)$ ,  $S_{XY}(\omega)$  and  $S_{YX}(\omega)$
- The power spectrum of  $W(t)$  as a function of  $S_{XX}(\omega)$ ,  $S_{YY}(\omega)$ ,  $\bar{X}$  and  $\bar{Y}$ , if  $X(t)$  and  $Y(t)$  are uncorrelated
- $S_{XW}(\omega)$  and  $S_{WX}$  as functions of  $S_{XX}(\omega)$ ,  $S_{YY}(\omega)$ ,  $S_{XY}(\omega)$  and  $S_{YX}(\omega)$

a)  $S_{ww}(w)$ ?

$$S_{ww}(w) = FT \langle R_{ww}(z) \rangle$$

$$R_{ww}(t, t+z) = E[w(t) \cdot w(t+z)] = E[(AX(t) + BY(t))(AX(t+z) + BY(t+z))]$$

$$= E[A^2 X(t)X(t+z)] + E[AB X(t)Y(t+z)] + E[AB Y(t)X(t+z)] + E[B^2 Y(t)Y(t+z)]$$

$$= A^2 R_{xx}(z) + AB R_{xy}(z) + AB R_{yx}(z) + B^2 R_{yy}(z) = R_{ww}(z)$$

$$S_{ww}(w) = FT \langle R_{ww}(z) \rangle = A^2 S_{xx}(w) + AB S_{xy}(w) + AB S_{yx}(w) + B^2 S_{yy}(w)$$

b)  $X(t)$  &  $Y(t)$  are uncorrelated  $\Rightarrow R_{xy}(t, t+z) = E[X(t)Y(t+z)] \stackrel{\text{uncorrelated}}{=} E[X(t)] E[Y(t+z)] \stackrel{w.c.c.}{=} \bar{X} \cdot \bar{Y}$

$$R_{ww}(z) = A^2 R_{xx}(z) + AB \underbrace{R_{xy}(z)}_{\bar{X} \cdot \bar{Y}} + AB \underbrace{R_{yx}(z)}_{\bar{X} \cdot \bar{Y}} + B^2 R_{yy}(z)$$

$$= A^2 R_{xx}(z) + 2AB \bar{X} \bar{Y} + B^2 R_{yy}(z)$$

$$S_{ww}(w) = FT \langle R_{ww}(z) \rangle = A^2 S_{xx}(w) + 2 \cdot 2AB \bar{X} \bar{Y} \underbrace{\int_{-\infty}^{\infty} \frac{d}{2\pi} d(w)}_{\text{Time}} + B^2 S_{yy}(w)$$

$$= A^2 S_{xx}(w) + B^2 S_{yy}(w) + 4\pi \cdot AB \bar{X} \bar{Y} \delta(w)$$



c) 1.  $S_{WX}(\omega)$ ?  
 2.  $S_{XW}(\omega)$ ?

1.  $S_{WX}(\omega) = FT \langle R_{WX}(z) \rangle$   
 $R_{WX}(z) = E[W(t) \cdot X(t+z)] = E[(AX(t) + BY(t)) \cdot X(t+z)]$   
 $= AR_{XX}(z) + BR_{YX}(z)$   
 $\rightarrow S_{WX}(\omega) = AS_{XX}(\omega) + BS_{YX}(\omega)$

2.  $S_{WX}(\omega) = \int (S_{XW}(\omega)) ? \rightarrow \dots$

Look for the relationship between  $S_{WX}(\omega)$  and  $S_{XW}(\omega)$ .

2.16 A wide-sense stationary  $X(t)$  is applied to an ideal differentiator, so that  $Y(t) = dX(t)/dt$ . The cross-correlation of the input-output processes is known to be

$$R_{XY}(\tau) = dR_{XX}(\tau)/d\tau$$

- (a) Determine  $S_{XY}(\omega)$  in terms of  $S_{XX}(\omega)$
- (b) Determine  $S_{YX}(\omega)$  in terms of  $S_{XX}(\omega)$

TIP: The key here is that the  $FT \left\{ \frac{df(\tau)}{d\tau} \right\} = j\omega \cdot FT\{f(\tau)\}$

2.17 The cross-correlation of jointly wide-sense stationary processes  $X(t)$  and  $Y(t)$  is assumed to be

$$R_{XY}(\tau) = Be^{-W\tau}u(\tau) \text{ where } B > 0 \text{ and } W > 0$$

are constants. Find

- (a)  $R_{YX}(\tau)$
- (b)  $S_{XY}(\omega)$  (use appendix C from Peebles' book)
- (c)  $S_{YX}(\omega)$  (use cross-power density properties)

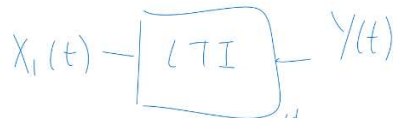
2.18 Consider two random process  $X_1(t)$  and  $X_2(t)$ . The mean of  $X_1(t)$  is equal to  $A$  ( $A > 0$ ) and  $X_2(t)$  is a white noise with power density 5

W/(rad/s). Given an LTI system with impulse response

$h(t) = e^{-\alpha}u(t)$ , with  $\alpha > 0$ . Find

- (a) The mean value of the response of the LTI system if the input is  $X_1(t)$
- (b) The average power (second-order moment) of the response of the system if the input is  $X_2(t)$ .

a)  $E[X_1(t)] = A > 0$



$h(t) = e^{-\alpha}u(t) = \bar{e}^{-\alpha t}, t \geq 0$

w.s.s.  $E[Y(t)] \rightarrow \bar{Y}?$

$$E[Y(t)] = \bar{X} \cdot \int_{-\infty}^{\infty} h(t) dt = \bar{X} \int_0^{\infty} e^{-\alpha t} dt$$

$$= \bar{X} \cdot \left( \frac{-1}{\alpha} \right) e^{-\alpha t} \Big|_0^{\infty} = \frac{\bar{X}}{\alpha}$$

b)

$X_2(t)$  (WHITE NOISE)  $\rightarrow$  [LTI]  $\rightarrow Y(t)$   $P_{YY}?$

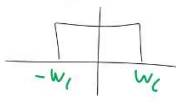
$$P_{YY} = \int_{u_2=-\infty}^{\infty} \int_{u_1=-\infty}^{\infty} R_{XX}(u_1-u_2) h(u_2) h(u_1) du_1 du_2$$

WHITE NOISE  $\downarrow$   $R_{XX}(\tau) = \frac{N_0}{2} \delta(\tau)$

$$P_{YY} = \frac{N_0}{2} \int_{-\infty}^{\infty} h^2(t) dt = \frac{N_0}{2} \int_0^{\infty} e^{-2\alpha t} dt = \frac{N_0}{4\alpha}$$

$$H(\omega) = \mathcal{FT}\{h(t)\} = \frac{1}{\alpha + j\omega} = \frac{1/\alpha}{1 + j\frac{\omega}{\alpha}}$$

$\omega_c = \alpha$



2.19 A random process  $X(t)$  with known mean  $X$  is the input of an LTI system with impulse response

$$h(t) = te^{-Wt}u(t).$$

Find

- The mean value of the response of the LTI system
- The average power (second-order moment) of the response of the system if  $X(t)$  is a white noise with power density  $5 W/(\text{rad/s})$ .

Quite similar to 2.18.

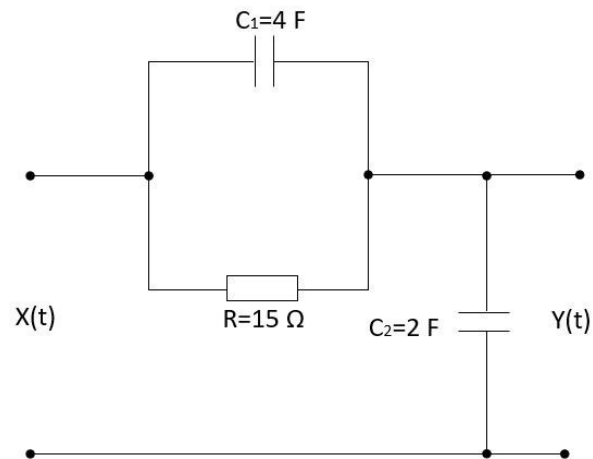
2.20 A white noise with power density  $N_0/2$  is applied to a network with impulse response of a system with impulse response

$$h(t) = Wte^{-Wt}u(t)$$

where  $W$  is a real positive constant. Find the cross-correlation of the response of input and the output of the system.

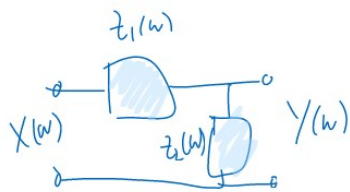
$$R_{XY}(\tau) = R_{XX}(\tau) * h(\tau) = \left[ \frac{N_0}{2} \delta(\tau) \right] * h(\tau) = \frac{N_0}{2} h(\tau) = \frac{WN_0}{2} \tau \cdot e^{-W\tau}, \tau \geq 0$$

- 2.21 A stationary random process  $X(t)$ , having an autocorrelation function  $R_{XX} = 2e^{-4|t|}$  is applied to the network of the figure below. Find the power spectrum of the output of the system.



$$S_{YY}(\omega) ?$$

$$S_{YY}(\omega) = S_{XX}(\omega) |H(\omega)|^2$$



$$\begin{aligned}z_{C_1}(w) &= \frac{1}{jwC_1} \\z_R(w) &= R \\z_{C_2}(w) &= \frac{1}{jwC_2}\end{aligned}$$

$$H(w) = \frac{z_2(w)}{z_1(w) + z_2(w)}$$

↙

$$z_1(w) = z_{C_1}(w) \parallel z_R(w) = \frac{z_{C_1}(w) z_R(w)}{z_{C_1}(w) + z_R(w)}$$
$$z_2(w) = z_{C_2}(w)$$

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$$Z_1(\omega) = \frac{\frac{1}{j\omega C_1} \times R}{\frac{1}{j\omega C_1} + R} = \frac{R}{1 + j\omega R C_1}$$

$$Z_2(\omega) = \frac{1}{j\omega C_2}$$

$$H(\omega) = \frac{Z_2(\omega)}{Z_1(\omega) + Z_2(\omega)} = \frac{\frac{1}{j\omega C_2}}{\frac{R}{1 + j\omega R C_1} + \frac{1}{j\omega C_2}} = \frac{1}{\frac{j\omega R C_2}{1 + j\omega R C_1} + 1}$$

$$= \frac{1 + j\omega R C_1}{j\omega R C_2 + 1 + j\omega R C_1} = \frac{1 + j\omega R C_1}{1 + j\omega R (C_1 + C_2)}$$

$\leftarrow$  ZEROS  
 $\leftarrow$  POLES

$$S_{YY}(\omega) = S_{XX}(\omega) \cdot |H(\omega)|^2 = S_{XX}(\omega) \frac{1 + (\omega R C_1)^2}{(1 + (\omega R (C_1 + C_2))^2)^2}$$

$$= \text{FT} \left\{ 2e^{-4|t|} \right\} \frac{1 + (\omega R C_1)^2}{(1 + (\omega R (C_1 + C_2))^2)^2}$$

$$= \underbrace{2 \frac{8}{16 + \omega^2}}_{16} \frac{1 + (\omega R C_1)^2}{1 + (\omega R (C_1 + C_2))^2}$$

- 2.22 A white noise  $X(t)$  with  $R_{XX} = 4 \cdot 10^{-3} \cdot \delta(\tau)$  is filtered with the network of the figure below. Find the average power of the input and output of the system.

